AMENDMENTS TO THE CLAIMS

This listing of claims replaces all prior versions, and listings, of claims in the application:

- 1 1. (Currently Amended) A method for detecting a signal burst transmitted on the initiative
- 2 of a sender on a radio channel listened to by a receiver system, the transmitted burst representing
- 3 a predetermined digital sequence, in which method channel parameters representing a statistical
- 4 behavior of the radio channel are estimated and a signal burst detection magnitude is evaluated
- 5 on the basis of the estimated channel parameters and of a correlation between a signal received at
- 6 the receiver system and the predetermined digital sequence, wherein said estimated channel
- 7 parameters comprise moments of order greater than 2 of the gain on the radio channel.
- 1 2. (Cancelled)
- 1 3. (Original) The method as claimed in claim 1, in which the signal received is subjected to
- 2 a filtering matched to the predetermined digital sequence so as to obtain said correlation in the
- 3 form of a complex signal having a first component on an in-phase path and a second component
- 4 on a quadrature path.
- 1 4. (Original) The method as claimed in claim 3, in which the evaluated detection magnitude
- 2 is proportional to $\left(\sum_{n=0}^{k} \frac{1}{n!(\sqrt{N_0})^n} \cdot H_n\left(\frac{z_x}{\sqrt{N_0}}\right) \cdot ma_{x,n}\right) \cdot \left(\sum_{n=0}^{k} \frac{1}{n!(\sqrt{N_0})^n} \cdot H_n\left(\frac{z_y}{\sqrt{N_0}}\right) \cdot ma_{y,n}\right)$, where
- 3 N_0 denotes the estimated power of the noise on the radio channel, z_x and z_y denote said first and
- 4 second components, $ma_{x,n}$ and $ma_{y,n}$ denote the moments of order n of the gain on the in-phase
- 5 path and on the quadrature path respectively, H_n denotes the Hermite polynomial of order n and
- 6 k is an integer larger than 2.

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- 1 5. (Original) The method as claimed in claim 1, in which said sender is a mobile terminal,
- 2 said receiver system belongs to a radiocommunication network and in which said burst is sent so
- 3 as to request access to the network.
- 1 6. (Original) The method as claimed in claim 1, in which said sender comprises a base
- 2 station of a radiocommunication network, said receiver system forms part of a mobile terminal,
- 3 and in which said burst is sent for the temporal synchronization between the sender and the
- 4 receiver system.
- 1 7. (Original) The method as claimed in claim 1, in which the detection of the burst is
- 2 utilized to select fingers of a rake receiver.
- 1 8. (Original) The method as claimed in claim 1, in which the burst belongs to a radio signal
- 2 sequence sent periodically, and in which said moments are estimated over a duration covering
- 3 several periods of said radio signal sequence.
- 1 9. (Currently Amended) A radio receiver system capable of detecting a signal burst
- 2 transmitted on the initiative of a sender on a radio channel listened to by the receiver system, the
- 3 transmitted burst representing a predetermined digital sequence, the receiver system comprising
- 4 means for estimating channel parameters representing a statistical behavior of the radio channel
- 5 and means for evaluating a <u>signal burst</u> detection magnitude on the basis of the estimated
- 6 channel parameters and of a correlation between a signal received at the receiver system and the
- 7 predetermined digital sequence, wherein said estimated channel parameters comprise moments
- 8 of order greater than 2 of the gain on the radio channel.
- 1 10. (Cancelled)

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- 1 11. (Original) A radio receiver system as claimed in claim 9, further comprising means for
- 2 subjecting the received signal to a filtering matched to the predetermined digital sequence so as
- 3 to obtain said correlation in the form of a complex signal having a first component on an in-
- 4 phase path and a second component on a quadrature path.
- 1 12. (Original) A radio receiver system as claimed in claim 11, in which the evaluated
- 2 detection magnitude is proportional to

$$3 \qquad \left(\sum_{n=0}^{k} \frac{1}{n!(\sqrt{N_0})^n} \cdot H_n\left(\frac{z_x}{\sqrt{N_0}}\right) \cdot ma_{x,n}\right) \cdot \left(\sum_{n=0}^{k} \frac{1}{n!(\sqrt{N_0})^n} \cdot H_n\left(\frac{z_y}{\sqrt{N_0}}\right) \cdot ma_{y,n}\right), \text{ where N}_0 \text{ denotes the}$$

- 4 estimated power of the noise on the radio channel, z_x and z_y denote said first and second
- 5 components, ma_{x,n} and ma_{y,n} denote the moments of order n of the gain on the in-phase path and
- on the quadrature path respectively, H_n denotes the Hermite polynomial of order n and k is an
- 7 integer larger than 2.
- 1 13. (Original) A radio receiver system as claimed in claim 9, belonging to a
- 2 radiocommunication network, said sender being a mobile terminal, and said burst being sent so
- 3 as to request access to the network.
- 1 14. (Original) A radio receiver system as claimed in claim 9, forming part of a mobile
- 2 terminal, said sender comprising a base station of a radiocommunication network, and said burst
- 3 being sent for the temporal synchronization between the sender and the receiver system.
- 1 15. (Original) A radio receiver system as claimed in claim 9, further comprising means for
- 2 utilizing the detection of the burst to select fingers of a rake receiver.

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- 1 16. (Original) radio receiver system as claimed in claim 9, in which the burst belongs to a
- 2 radio signal sequence sent periodically, and in which said moments are estimated over a duration
- 3 covering several periods of said radio signal sequence.